# Three Discretizations of the MFIE for the Linear Dipole 

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## 1. Introduction

Procedures for determining the currents and input impedance of a linear dipole antenna are generally based on studies of the electric-field integral equation (EFIE), specifically the Pocklington or Hallén equations. Approaches based on the magnetic field integral equation (MFIE) have received little attention other than the work of Zhang and his colleagues in the 1980s [1-3]. Zhang's MFIE implementation employed a series feed and used a low-order method of moments discretization. In the following, the MFIE with a frill feed model is considered. Three discretization procedures, the method of moments (MoM), the boundary residual method (BRM) [4-5], and the locally-corrected Nyström (LCN) method [6-7], are investigated. Since the MFIE is only rigorously applicable to closed conducting bodies, currents on flat end caps are included in the model.

## 2. Formulation

Consider a linear dipole, of radius $a$, aligned with the $z$-axis in a cylindrical coordinate system. By assumption, the fields and current density are $\phi$ symmetric. The MFIE for the linear dipole is based on the general relation

$$
\begin{equation*}
H_{\phi}^{\text {inc }}=H_{\phi}^{\text {tot }}-H_{\phi}^{s} \tag{1}
\end{equation*}
$$

which will be enforced in the limit as the observer approaches the surface of the cylindrical dipole from the exterior. The primary unknown on the cylindrical part of the dipole can be expressed in terms of the total current $I(z)$ through the equation

$$
\begin{equation*}
I(z)=2 \pi a J_{z}(a, \phi, z) \tag{2}
\end{equation*}
$$

The scattered field due to currents on the barrel can be expressed as

$$
\begin{equation*}
H_{\phi}^{s}(\rho, 0, z)=\frac{1}{2 \pi} \int_{z} \int_{\phi} I\left(z^{\prime}\right) \frac{e^{-j k R}}{4 \pi R^{3}}(1+j k R)\left(\rho-a \cos \phi^{\prime}\right) d \phi^{\prime} d z^{\prime} \tag{3}
\end{equation*}
$$

where

$$
\begin{equation*}
R=\sqrt{\rho^{2}-2 \rho a \cos \phi^{\prime}+a^{2}+\left(z-z^{\prime}\right)^{2}} \tag{4}
\end{equation*}
$$

The scattered H -field due to a $\rho$-directed current density on an end cap of the dipole, expressed as $I(\rho)=2 \pi \rho J_{\rho}\left(\rho, \phi, z^{\prime}\right)$, can be written as

$$
\begin{equation*}
H_{\phi}^{s}(\rho, 0, z)=-\frac{1}{2 \pi} \int_{\rho} \int_{\phi} \cos \left(\phi^{\prime}\right) I\left(\rho^{\prime}\right) \frac{e^{-j k R}}{4 \pi R^{3}}(1+j k R)\left(z-z^{\prime}\right) d \rho^{\prime} d \phi^{\prime} \tag{5}
\end{equation*}
$$

where the end cap is located at $z^{\prime}$ and where

$$
\begin{equation*}
R=\sqrt{\rho^{2}-2 \rho \rho^{\prime} \cos \phi^{\prime}+\left(\rho^{\prime}\right)^{2}+\left(z-z^{\prime}\right)^{2}} \tag{6}
\end{equation*}
$$

The incident field is that produced by a magnetic frill of outer radius $b$ located at the center of the dipole, and can be expressed as

$$
\begin{equation*}
H_{\phi}^{i n c}(\rho, 0, z)=\frac{j \omega \varepsilon}{\ln (b / a)} \int_{\rho^{\prime}=a}^{b} \int_{\phi^{\prime}=0}^{2 \pi} \cos \left(\phi^{\prime}\right) \frac{e^{-j k R}}{4 \pi R} d \phi^{\prime} d \rho^{\prime} \tag{7}
\end{equation*}
$$

where $R$ is defined in Equation 6. These integrals can be computed by quadrature with appropriate treatment of the singularities at $R=0$.

## 3. Discretization Methods

The MoM procedure is well-known to the EM community and needs no elaboration. A discretization of the MFIE incorporating piecewise-linear basis functions was implemented, with the equation enforced at discrete points at cell junctions to produce a square system of equations.

The BRM, as described in [4-5], was also applied to the MFIE with piecewiselinear basis functions. In this BRM implementation, the equation is enforced at twice the number of boundary points as there are cells in the model, to yield an overdetermined system of equations. In an attempt to better model the current and charge density at the corners where the barrel of the dipole meets the endcaps, the basis functions adjacent to the function at the corner were modified to incorporate a functional dependence

$$
\begin{equation*}
B(u)=u^{2 / 3} \tag{8}
\end{equation*}
$$

in the cells abutting a corner, where $u$ is a parametric variable along either z or $\rho$ with an origin at the corner. Equation (8) is used instead of the linear function in order to provide the appropriate charge singularity. These functions are superimposed with the regular linear basis functions centered at the corners, so the current at the corner is nonzero.

The LCN procedure [6-7] was implemented with Gauss-Legendre rules of order $p$ up to 10 . Order $p=2$ is expected to provide a representation similar to that of the linear basis functions used with the MoM and BRM [7]. Local corrections were performed in the source cell and the adjacent cells. No attempt was made to incorporate the edge condition as in Equation (8). The LCN procedure does not impose cell-cell continuity and thus uses more unknowns for a given number of cells than the MoM or BRM approaches.

## 4. Results

For illustration, input impedance results are shown in Tables 1-5 for a dipole of length $0.48 \lambda$ and radius $\mathrm{a}=0.0391 \lambda$, for a frill feed corresponding to a coaxial ratio $\mathrm{b} / \mathrm{a}=1.187$. For this dipole, Holly measured an equivalent impedance of $\mathrm{Z}_{\mathrm{in}}=102$ - j $16 \Omega$ [8]. Numerical results for impedance from the three methods exhibit little variation with changes in the number of cells in the model and agree to within about $1 \Omega$ with each other. The low-order LCN method ( $p=2$ ) provides a representation comparable to the linear basis functions used with the MoM and BRM methods. Higher-order LCN results are shown in Tables 4 and 5, and show only a slight variation from the $p=2$ results. The total number of equations and unknowns are also shown for each approach.

## 5. References

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[4] K. R. Bunch and R. W. Grow, "Numerical aspects of the boundary residual method," Int. J. Num. Modelling, vol. 3, pp. 57-71, March 1990.
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[7] A. F. Peterson, "Accuracy of currents produced by the locally-corrected Nyström method and the method of moments when used with higher-order representations," ACES Journal, vol. 17, pp. 74-83, 2002.
[8] S. Holly, "Experimental study of electrically thick monopole antennas," Cruft lab., Harvard University, Cambridge MA, Sci. Rep. 8, Dec. 1969.

| TABLE 1. |  |  |  |  |  |  | MoM results for $\mathrm{Z}_{\text {in }}$ of a dipole with $\mathrm{L}=0.48 \lambda, \mathrm{a}=0.0391 \lambda, \mathrm{~b} / \mathrm{a}=1.187$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| cells on <br> barrel | cells on each <br> end | unknowns | equations | $\mathrm{R}_{\text {in }}$ | $\mathrm{X}_{\text {in }}$ |  |  |
| 26 | 1 | 27 | 27 | 83.32 | -11.63 |  |  |
| 50 | 2 | 53 | 53 | 92.00 | -13.59 |  |  |
| 74 | 3 | 79 | 79 | 94.29 | -14.11 |  |  |
| 98 | 4 | 105 | 105 | 95.23 | -14.31 |  |  |
| 124 | 5 | 133 | 133 | 95.73 | -14.41 |  |  |
| 148 | 6 | 159 | 159 | 96.00 | -14.47 |  |  |
| 172 | 7 | 185 | 185 | 96.18 | -14.50 |  |  |
| 196 | 8 | 211 | 211 | 96.30 | -14.52 |  |  |
| 222 | 9 | 239 | 239 | 96.39 | -14.54 |  |  |
| 246 | 10 | 265 | 265 | 96.45 | -14.55 |  |  |

TABLE 2. BRM results for a dipole with $\mathrm{L}=0.48 \lambda, \mathrm{a}=0.0391 \lambda, \mathrm{~b} / \mathrm{a}=1.187$

| cells on <br> barrel | cells on each <br> end | unknowns | equations | $\mathrm{R}_{\text {in }}$ | $\mathrm{X}_{\text {in }}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 26 | 1 | 27 | 56 | 104.52 | -11.85 |
| 50 | 2 | 53 | 108 | 98.24 | -13.87 |
| 74 | 3 | 79 | 160 | 95.97 | -13.79 |
| 98 | 4 | 105 | 212 | 95.90 | -14.34 |
| 124 | 5 | 133 | 268 | 95.82 | -14.57 |
| 148 | 6 | 159 | 320 | 95.76 | -14.65 |
| 172 | 7 | 185 | 372 | 95.82 | -14.74 |
| 196 | 8 | 211 | 424 | 95.84 | -14.77 |
| 222 | 9 | 239 | 480 | 95.87 | -14.81 |
| 246 | 10 | 265 | 532 | 95.89 | -14.83 |


| TABLE 3. LCN results for $\mathrm{p}=2$; dipole with $\mathrm{L}=0.48 \lambda, \mathrm{a}=0.0391 \lambda, \mathrm{~b} / \mathrm{a}=1.187$ |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| cells on <br> barrel | cells on each <br> end | unknowns | equations | $\mathrm{R}_{\text {in }}$ | $\mathrm{X}_{\text {in }}$ |  |
| 26 | 1 | 56 | 56 | 99.74 | -4.90 |  |
| 50 | 2 | 108 | 108 | 98.45 | -9.25 |  |
| 74 | 3 | 160 | 160 | 97.96 | -10.92 |  |
| 98 | 4 | 212 | 212 | 97.71 | -11.79 |  |
| 124 | 5 | 268 | 268 | 97.57 | -12.35 |  |
| 148 | 6 | 320 | 320 | 97.49 | -12.69 |  |
| 172 | 7 | 372 | 372 | 97.41 | -12.95 |  |
| 196 | 8 | 424 | 424 | 97.34 | -13.15 |  |
| 222 | 9 | 480 | 480 | 97.26 | -13.33 |  |
| 246 | 10 | 532 | 532 | 97.18 | -13.47 |  |

TABLE 4. LCN results for $\mathrm{p}=6$; dipole with $\mathrm{L}=0.48 \lambda, \mathrm{a}=0.0391 \lambda, \mathrm{~b} / \mathrm{a}=1.187$

| cells on <br> barrel | cells on each <br> end | unknowns | equations | $\mathrm{R}_{\text {in }}$ | $\mathrm{X}_{\text {in }}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 12 | 1 | 84 | 84 | 97.88 | -10.96 |
| 24 | 2 | 168 | 168 | 97.40 | -12.74 |
| 36 | 3 | 252 | 252 | 97.22 | -13.35 |
| 48 | 4 | 336 | 336 | 97.12 | -13.67 |
| 60 | 5 | 420 | 420 | 97.06 | -13.87 |
| 72 | 6 | 504 | 504 | 97.02 | -14.00 |

TABLE 5. $L C N$ results for $\mathrm{p}=9$; dipole with $\mathrm{L}=0.48 \lambda, \mathrm{a}=0.0391 \lambda, \mathrm{~b} / \mathrm{a}=1.187$

| cells on <br> barrel | cells on each <br> end | unknowns | equations | $\mathrm{R}_{\text {in }}$ | $\mathrm{X}_{\text {in }}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 12 | 1 | 126 | 126 | 97.37 | -12.84 |
| 24 | 2 | 252 | 252 | 97.11 | -13.72 |
| 36 | 3 | 378 | 378 | 97.02 | -14.02 |
| 50 | 4 | 522 | 522 | 96.98 | -14.19 |

